# REAL TIME ESTIMATION OF MODAL PARAMETERS OF NON-STATIONARY SYSTEMS USING ADAPTIVE WAVELET FILTERING AND RECURSIVE LEAST SQUARE ALGORITHM

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#### Summary

The paper presents a method of modal parameter estimation based on RLS (Recursive Least Square) algorithm, and wavelet filtering. The wavelet filtering gives possibility to decoupling frequency components of signal response of structure. This operation can also reduce the order of the signal model estimated by RLS algorithm. An additional advantage of this method is the possibility of adapting the wavelet filter parameters to the changing parameters of the system. Reduced model order significantly reduces the time of estimation of modal parameters, which enables the real – time implementation of the method. Due to recursively updated covariance matrix of model parameters, the confidence intervals of modal parameters can be also estimated. All routines have been implemented and tested in MATLAB®. The method have been tested on simulated data delivered by an AIRBUS team and on the test bed with a variable stiffness.

Keywords: Wavelet transform, recursive identification, non-stationary systems.

# IDENTYFIKACJA PARAMETRÓW MODALNYCH UKŁADÓW NIESTACJONARNYCH Z WYKORZYSTANIEM ADAPTACYJNEGO FILTRU FALKOWEGO ORAZ REKURSYWNYEGO ALGORYTMU NAJMNIEJSZYCH KWADRATÓW

### Streszczenie

Artykuł prezentuje metodę estymacji parametrów modalnych bazująca na algorytmie RLS (RLS (Recursive Least Square) oraz filtracji falkowej. Filtracja falkowa daje możliwość separacji składników częstotliwościowych sygnału. Ta operacja redukuje rząd modelu estymowanego przez algorytm RLS. Dodatkowa zaletą algorytmu jest możliwość adaptacji parametrów filtru falkowego do zmieniających się parametrów układu. Redukcja modelu znacznie skraca czas estymacji parametrów modalnych. Umożliwia to implementację algorytmu w czasie rzeczywistym. Dzięki rekursywnemu uaktualnianiu macierzy kowariancji parametrów modelu estymowane są również przedziały ufności otrzymanych wyników. Wszystkie procedury zostały zaimplementowane w środowisku MATLAB. Metodę przetestowano dla danych symulacyjnych (model samolotu dostarczony przez firmę AIRBUS), oraz dla układu ze zmienna sztywnością.

Słowa kluczowe: Transformata falkowa, rekursywne metody identyfikacji, układy niestacjonarne.

# 1. INTRODUCTION

Many practical engineering systems change dynamic parameters during their operation. Possible reason of parameters change can be damage, changes of operational conditions or occur some physical phenomena causing changes in dynamics of the system. Analysis of nonstationary systems are more difficult than in stationary case and require a dedicated method of identification.

Nonstationary linear mechanical systems are those systems whose parameters as stiffness, natural frequency, mass and damping ratios or statistical properties of an input signal (such as mean value and variance) are time dependent. Changes of this parameters can be caused by many factors, from changes of environmental conditions like temperature, humidity, changes of structure geometry, variable operational conditions to damage. In practice, many engineering structures like traffic - excited bridges, rotating machinery working with varying speed, aircrafts, robots, cranes and many others should be treated as non stationary systems. Analysis of non - stationary systems require application of dedicated method which can not be based on averaging which is most popular way of improving quality of parameters identification results. In some cases, when changes of parameters are slow, the system can be treated as stationary in given time interval. Firsts attempts of non - stationary systems parameters identification consists in adaptation of classical identification method of Linear Time Invariant (LTI) systems. Main problem of this approach was to adapt classical algorithm to operate on small number of samples, where quasistationarity condition was assumed. Within the space of the last several dozen years many method for non - stationary system identification were created. The method based on different algorithms like for example BTLS (Bootstarpped Total Least Squares) RLS (Recursive Least Square), TARMA (Timedependent Autoregressive Moving Average). Recursive Subspace and many others. In the literature many of application of this method can be found.

One of an example of phenomenon causing nonstationary behavior of aircrafts (but not only) can be flutter phenomenon. Unstable vibrations of an airplane can be a reason of a catastrophic failure of the aircraft [1]. A critical instability phenomenon is known as "aero-elastic flutter". In the literature [2-4] many cases of flutter phenomena are carefully studied. The importance of accurate definition aeroelastic effects such as flutter on flight vehicles are evident and in fact can be traced earliest days of manned aviation itself. For preventing from a flutter phenomenon, the airplane is submitted to a flight flutter testing procedure, with incrementally increasing altitude and airspeed. Flight flutter testing procedure can be formulated as procedure of stability test. One possible solution is to find instable poles (which are responsible of flutter phenomena) of the aircraft structure with employing modal model parameters. Important challenges of the in-flight modal analyses are the limited choices for measured excitation inputs, and the presence of unmeasured natural excitation input (e.g. turbulence). A better exploitation of flight test data can be achieved by using output-only system identification methods, which exploits recorded vibration data under natural excitation conditions, without artificial control surface excitation and other types of excitation inputs [1,3]. There are many different modal parameters identification methods that could be used for flight flutter testing [4-7].

This paper presents a recursive modal parameters identification method of nonstationary systems. The classical RLS algorithm is supported wavelet transform which allows decouple components of the signal response and reduce the order of the model. This operation greatly accelerated the process of modal parameter estimation, which is a critical part of the algorithm. Such an approach allow real – time implementation of the algorithm. The method was applied to identification of modal parameters of airplane model and to track frequency changes of the system with variable stiffness.

# 2. RECURSIVE METHOD OF MODAL PARAMETERS ESTIMATION

The proposed algorithm consists of three main parts. In the first step the signal is decomposed by wavelet transform. In second step model parameters are estimated. In third step a modal parameters and their standard deviation are determined. Organization of the algorithm shown in Figure 1.



# 2.1 Wavelet Transform

The wavelet analysis is a method of signal decomposition. As a result of the wavelet analysis, in contradiction to the Fourier transform, elementary signals – so called wavelets – are obtained. Wavelet functions are continuous, oscillated with various duration times and spectrums. From the mathematical point of view, a continuous wavelet transform (CWT) of a signal x(t) can be defined as [8,9]

$$\left(W_{g}x\right)(a,b) = \frac{1}{\sqrt{a}} \int_{-\infty}^{+\infty} x(t)g^{*}\left(\frac{t-b}{a}\right) dt$$
(1)

Using properties of wavelet transform, can be proved mathematically that this kind of time frequency analysis decoupling frequency components of signal. Utilizing above - mentioned feature as signal processing method each component of the signal can be analyzed separately. Additionally order of the model using to identification process is reduced to model for only given frequency component of the signal. This approach decrease computational effort which can be significant for higher model order systems especially during finding a roots of the characteristic polynomial.



Fig. 2. Wavelet based signal components decoupling

Besides, it is not always demand to track all natural frequencies and damping ratios of the system. Using wavelet filtration only critical modes (from the point of view of process) can be track. It also reduces demand for computational capability which increase applicability of method and makes real – time implementation process easier. Schematically process of frequency component decoupling is shown in Figure 2. More detailed information about frequency components decoupling can be found at [9].

### 2.2. Recursive Least Square algorithm

Method of estimation model coefficients based on RLS algorithm. Schematically the algorithm is presented in Figure 3.



Fig. 3. Organization of Recursive Least Square method

The *i* indicate successive number of sample from A/D converter or memory buffer. Determination of modal parameters from the received signal model is time consuming for of higher orders models and require to find the roots of the characteristic polynomials. By using wavelet filtering, only one frequency component of the signal is analyzed. This determines the order of estimating model. It is worth taking into consideration that by using wavelet transform the problem of selection the model order is solved. This problem (though very important) will not be discussed in this article.

# 2.3. Modal parameters and their standard deviation

Thanks to low (equal to 2) model order, estimation of modal parameters is reduced to the application of simple mathematical formulas. For example, the natural frequency of the system and the corresponding damping coefficient can be determined from the dependence:

$$p = \frac{1}{2} \frac{\sqrt{\ln\left(\frac{1}{2}\sqrt{a_1 + |a_1 - 4a_2|}\right)}}{T_s} + \frac{\arctan\left(\frac{1}{2}\sqrt{a_1 + |a_1 - 4a_2|}\right)}{T_s}$$
(2)

$$\begin{split} \zeta &= -\frac{\ln \! \left( \frac{1}{2} \sqrt{a_1^2 + \left| a_1^2 - 4a_2 \right|} \right)}{T_z \sqrt{\frac{\ln \! \left( \frac{1}{2} \sqrt{a_1^2 + \left| a_1^2 - 4a_2 \right|} \right)^2}{T_z^2}} + \frac{\arctan \! \left( \frac{1}{2} \sqrt{\left| a_1^2 - 4a_2 \right|} , - \frac{1}{2} a_1 \right)^2}}{T_z^2} \end{split}$$

where:  $a_1$ ,  $a_2$  – model coefficients,  $\omega$  - natural frequency,  $\zeta$  - damping ratio,  $T_s$  – sampling time.

When analytical dependences between modal and model parameters are known it is possible to calculate covariance matrix of modal parameters [10,11].

$$\hat{P}_{\kappa}(\hat{\kappa}_{n}) = E\left[\left(\kappa_{0} - \hat{\kappa}_{N}\right)\left(\kappa_{0} - \hat{\kappa}_{N}\right)^{T}\right]$$
(3)

where: E - is the expected value,  $\hat{\kappa}_N = [\omega_1\zeta_1, \omega_2\zeta_2, ..., \omega_s, \zeta_n]$  - is a vector of estimated modal parameters,  $\kappa_0$  - is a vector of true modal parameters,  $\hat{P}_{\kappa}(\hat{\kappa}_n)$  - is a covariance matrix of modal parameters. Using Taylor series expansion method, the covariance matrix of modal parameters can be expressed as:

$$\hat{P}_{\kappa}(\hat{\kappa}_{n}) = J(\hat{\theta}_{N})E\Big[\Big(\theta_{0} - \hat{\theta}_{N}\Big)(\theta_{0} - \hat{\theta}_{N}\Big)^{T}\Big]J(\hat{\theta}_{N})^{T} = J(\hat{\theta}_{N})P_{\theta}(\hat{\theta}_{n})J(\hat{\theta}_{N})^{T} \quad (4)$$

Where  $J(\hat{\theta}_N)$  is the following Jacobi matrix:

$$J(\theta) = \begin{bmatrix} \frac{\partial f_1(\theta)}{\partial \theta_1} & \frac{\partial f_1(\theta)}{\partial \theta_2} & \dots & \frac{\partial f_1(\theta)}{\partial \theta_n} \\ \frac{\partial f_2(\theta)}{\partial \theta_1} & \frac{\partial f_2(\theta)}{\partial \theta_2} & \dots & \frac{\partial f_2(\theta)}{\partial \theta_n} \\ \vdots & \vdots & \dots & \vdots \\ \frac{\partial f_m(\theta)}{\partial \theta_1} & \frac{\partial f_m(\theta)}{\partial \theta_2} & \dots & \frac{\partial f_m(\theta)}{\partial \theta_n} \end{bmatrix}$$
(5)

and  $\theta$  - is the vector of model parameters. In general case the Jacobi matrix is impossible to determine, that is why in the considered case the numerical differentiation with the Central Difference method was applied. When the standard deviation of modal parameters is known, the confidence bounds can be easy determined using "*n* sigma" method. In this work "three sigma" was assumed. It means that about 99,7% of the values lie within three standard deviation range from expected value.

## 3. ADAPTIVE WAVELET FILTERING

Main difficulty of the identification process with using wavelet transform algorithm is selection of parameters responsible for decoupling of signal frequency components. The issue connected with proper wavelet function parameters selection mainly concern the Heisenberg relation. According to this rule the signal cannot be analyzing with the same good resolution in time and frequency domain simultaneously. Selection of wavelet function require some compromise between quality of information from frequency and time domain. Additional inconvenience of using constant wavelet function for identification process is that it has limited bandwidth. In situation where changes of signal frequency component are grater then half of assumed bandwidth parameter (to assume that central frequency of wavelet filter is tuned to frequency of given component) filtration process can be incorrect and obtained results can be only filter response. This problem is presented in figure 2a. Detailed information about non - adaptive wavelet filtering method for system parameter identification can be found at (Klepka A. Uhl T 2008, Uhl T. 2008). Solution of constant filter bandwidth problem can be make bandwidth parameter conditional on process. This identification requires the determination of wavelet functions for different moments in time. Applying that, formula 1 must be rewritten as:

$$\left(W_{g_i}x_i\right)\left(a_i,b_i\right) = \frac{1}{\sqrt{a_i}} \int_{-\infty}^{+\infty} x_i(t)g_i^*\left(\frac{t_i-b_i}{a_i}\right) dt \quad (6)$$

where i is given time interval. Determination of wavelet functions which allows filtration of the given frequency component requires definition of a scale parameter associated with the frequency by the formula:

$$f_i = \frac{T_s}{a_i} \tag{7}$$

where  $T_s$  is sampling time and  $f_i$  is frequency corresponding to scale parameters  $a_i$ . From the above relation shows that if the scale parameter changes will also change the frequency of the signal filtering. This allows to change the band wavelet filter during identification process.

In the presented algorithm, the scale parameter  $a_i$  is determined based on the frequency  $f_i$  estimated from recursive model.



Fig. 4. Scheme of adaptive procedure organization

Adaptation process is realized by comparing wavelet frequency and frequency obtained from RLS algorithm. If the frequencies have the same value or their difference contain in given range, identification process is continued without any changes, but if this two frequencies have different value or their difference has value greater than assumed, the frequency of wavelet function is changed to value corresponding to frequency value estimated from RLS algorithm. Schematically, process adaptation and diagram of method with adaptive wavelet filtering is presented in figure 4, where  $f_e$  and  $f_w$  are current estimated frequency and wavelet frequency respectively.

### 4. CASE STUDIES

### 4.1. An aero – elastic model

The aeroelastic equations define the time evolution of the vector of the structure generalized displacements by the second-order differential equation [12]. The left-hand side of this equation concerns the structural efforts while the right-hand side is the sum of various external forces. The vector of the measurements depends linearly on and its first and second derivatives according to equation :

$$M\ddot{\mathbf{q}}(t) + (A + G)\dot{\mathbf{q}}(t) + K\mathbf{q}(t) = F_a(t) + F_c(t) + F_t(t) \quad (8)$$

q(t) - Vector of the generalized displacements, M, A, K - Structural mass, damping and stiffness, G - Gyroscopic terms due to the engines,  $F_a(t)$  - Unsteady aerodynamic forces,  $F_c(t)$  - Control surface forces function of the deflections  $\delta(t)$ ,  $F_t(t)$  - Turbulence forces function of the wind speeds W(t). Based on presented model the simulation was perform. The scenario of simulation assumed that speed of plane would be uniformly increased form 330 kts to 360 kts.

The results of modal parameters identification of the model for mode responsible for flutter phenomena shown in Figures 5 and 6.



Fig. 5. Comparison of identification results: a) damping ratio, b) frequency



Fig. 6. Identified time history of modal parameters non – stationary aero - elastic system with estimated confidence bounds: a) damping ratio, b) natural frequency

As can be noticed some peaks appeared in estimated confidence bounds. This occurrence is caused by wavelet adaptation process. In the next step after wavelet function changing, the RLS algorithm is called with initial parameters connected with "old" wavelet function. As a result of it covariance matrix values increase temporary. This problem was solved in hardware version where parallel computation can be perform.

### 4.2. System with variable stiffness

The hardware implementation of algorithm was tested on the 1DOF system with variable stiffness. Experimental setup of experiment is presents in Figure 7.



Fig. 7. Experiment arrangement

An electromagnetic shaker and signal generator were used to excite the structure. White noise signal

was used as excitation signal. During experiment, pressure in metal bellows was changed. As result of system stiffness changes natural frequency of the system was shifted.. Time history of system response, result of natural frequency identification are presented in figure 8.



Fig. 8. System response (a), Comparison of identified natural frequency of the system (b)

Additionally, wavelet adaptation process was presented in Figure 8b (red dashed line).

### 5. CONCLUSIONS

All performed test showed that adaptive wavelet formula combined with RLS algorithm gives satisfactory results. Natural frequency of the systems are identified much better then damping ratios. Good results of natural frequency identification gives a possibility of use this kind of algorithm in other applications. An example can be vibrating string sensors where frequency of oscillation of sensor resonator can give information about changes of operational condition. Another example can be process of rotational speed tracking in structures like wind turbine, gear boxes or mining machines. A formulated algorithm allows computing modal parameters of structures in real - time. Hardware implementation of the algorithm is proposed with the Hardware-Software Co-design approach, i.e. a part realized by hardware and the remaining part by software running on a Nios II soft-processor contained in the FPGA. Using adaptive formula of wavelet filter, process of wavelet function selection was simplified. The method enable to track modal parameters of the non - stationary system even if their changes are significant.

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